Drude Model of Electronic Conduction 1

Simplest model: electrons are accelerated by applied \vec{E} and are colliding with scatterers. collisions are instantaneous processes when electrons change their \vec{v} . Thermalization occurs only during collisions, after collision \vec{v} corresponds to T.

Current density:
$$\vec{J}_e = -ne\langle \vec{v} \rangle$$
 Drift velocity of e⁻ gas: $m\vec{a} = \frac{d\vec{v}}{dt} = -e\vec{E} \Rightarrow \vec{v}(t) = -\frac{e\vec{E}}{m}t + \vec{v}(0)$

Collisions restore e- \vec{v} to local thermal equilibrium. We label t as elapsed time before previous collision. Is velocity at t_0 will be v_0 , and it will get some eE/m in between collisions. We assume random motion without contribution of v_0 to average electronic velocity.

Therefore:
$$\langle \vec{v} \rangle = -\frac{e\vec{E}}{m} \langle t \rangle = -\frac{e\tau}{m} \vec{E} \Rightarrow \vec{J} = \frac{ne^2\tau}{m} \vec{E} = \sigma\vec{E}$$

Defining mean free path
$$l_e = \langle v \rangle \tau$$
 we get: $\sigma = \frac{ne^2\tau}{m} = \frac{ne^2l_e}{m\langle v \rangle}$ and $\mu = \frac{e\tau}{m} = \frac{el_e}{m\langle v \rangle}$
Electron density: $n = N_A z \rho_m^{\uparrow} / A_{A \text{tomic weight}}$
Introduction to thermal and transport techniques

Drude Model of Electronic Conduction 2

In the absence of scattering $\tau \to \infty$ and therefore $\sigma \to \infty$. Real materials have defects, so there is always some residual scattering ~ τ_0 and σ_0 . We can assume that scattering from phonons is independent process, therefore we can write: Matthiessen's rule

$$\rho = \frac{m}{ne^2} \left(\frac{1}{\tau_0} + \frac{1}{\tau_{ph}} \right)$$

Time dependent electric field: $\vec{E}(t) = \vec{E}(\omega)e^{-i\omega t} \Rightarrow \sigma(\omega) = \frac{(ne^2\tau)/m}{1-i\omega\tau}$ Also using Maxwell equations we get wave equation: $\nabla^2 \vec{E} = -\frac{\omega^2}{c^2} \varepsilon(\omega) \vec{E}$ $\sigma_{_0}$

Where: $\varepsilon(\omega) = 1 + \frac{4\pi i \sigma}{\omega}$

For
$$\omega \tau >> 1$$
 $\varepsilon(\omega) = 1 - \frac{\omega_p^2}{\omega^2}; \omega_p^2 = \frac{4\pi n e^2}{m}$

 \rightarrow for $\omega < \omega_p$ electric field decays exponentially in electron gas

See derivation Aschroft-Mermin

 \rightarrow for $\omega > \omega_p$ solutions are oscillatory and radiation can propagate

Electrical Conductivity and Fermi-Dirac Statistics

Occupation probability

$$\vec{j} = -\frac{e}{4\pi^3} \int_{k_{osc}} v(\vec{k}) d\vec{k} = -\frac{e}{4\pi^3} \int_{1 \le BZ} v(\vec{k}) f(\vec{k}) d\vec{k}$$
due to scattering

$$f_{\sigma} = [\exp(a(\mathbf{k}) - \mu) + 1]^{-1}$$

$$\vec{j} = -\frac{e}{4\pi^3} \int_{k_{osc}} v(\vec{k}) d\vec{k} = -\frac{e}{4\pi^3} \int_{1 \le BZ} v(\vec{k}) f(\vec{k}) d\vec{k}$$
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$$\vec{j} = -\frac{e}{4\pi^3} v(\vec{k}) d\vec{k}$$

$$\vec{j} = -\frac{e}{4\pi$$

$$\vec{j} \approx -\frac{e}{4\pi^3} \int v(\vec{k}) \left[f_0(\vec{k}) + \frac{e}{\hbar} \tau(\vec{k}) \vec{E} \nabla_k f_0(\vec{k}) \right] d\vec{k} \qquad dk = dS_e dk_\perp = dS_e \frac{de}{|\nabla_k \varepsilon|} = dS_e \frac{de}{\hbar v(\vec{k})}$$

$$\Rightarrow j_x - \frac{e^2}{4\pi^3} \int v_x(\vec{k}) \left[f_0(\vec{k}) + \frac{e}{\hbar} \tau(\vec{k}) E_x \frac{\partial f_0}{\partial x} \right] d\vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right] \vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right) \vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right) \vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right) \vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right) \vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right) \vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right) \vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right) \vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right) \vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right) \vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right) \vec{k} = \left(\frac{e^2 E_x}{4\pi^3} \int v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial \varepsilon} d\vec{k} \right) \vec{k}$$

Introduction to thermal and transport techniques

Since $\partial f_0 / \partial k_x = \hbar v_x (\partial f_0 / \partial \epsilon)$

Resistance and Resistivity in Experiment

 $\begin{array}{ll} \text{Resistivity in Drude model} & \rho(T) \text{ is determined by } \tau(T): \\ e^- \text{ ph} (T) \text{ is determined by } \tau(T): \\ e^- \text{ ph} (T) \text{ is determined by } \tau(T): \\ e^- \text{ ph} (T) \text{ is determined by } \tau(T): \\ e^- \text{ ph} (T) \text{ ph}^{-1} T \rightarrow \rho_{\text{ph}} \sim T \\ e^- \text{ ph} (T) \text{ ph}^{-1} T^3 \text{ but } \rho_{\text{ph}} \sim T^5 (\text{Bloch-Grüneisen}) \text{ due to small angle scattering e}^{-1} \\ e^- \text{ ph} (T) \text{ ph}^{-1} T^3 \text{ but } \rho_{\text{ph}} \sim T^5 (\text{Bloch-Grüneisen}) \text{ due to small angle scattering e}^{-1} \\ e^- \text{ at low t} \rho \sim T^2 (\text{Fermi Liquid}), \text{ or } \rho \sim T (\text{non-Fermi liquid}), \text{ etc...}) \\ \text{Therefore } \rho(T, H) \text{ can tell us something about interactions - but - one has to note all (scattering) processes in order to be able to analyze the data.} \end{array}$

Which gives us some basic insight into electronic transport in new material: n and / or τ

$$\sigma \approx \frac{e^2}{4\pi^3\hbar} \int \frac{v_x^2(\vec{k})}{v(\vec{k})} \tau(\vec{k}) \delta(\varepsilon - \varepsilon_F) dS_{\varepsilon} d\varepsilon = \frac{e^2}{4\pi^3\hbar} \int_{\varepsilon = \varepsilon_F} \frac{v_x^2(\vec{k})}{v(\vec{k})} \tau(\vec{k}) dS_{\varepsilon}$$

Which tells us that electrical conductivity of a metal is a surface integral over the Fermi surface in k space – it is **proportional to the area of Fermi surface available for conduction**.

NOTE: the measured electrical resistivity will contain an impurity scattering term, ρ_0 which appears additively. This is often associated with chemical impurities as well as a variety of structural defects.



Four Wire DC and AC Resistance Measurement

Two contacts are used to pass a current through the sample, two leads are used in order to measure the voltage drop. This configuration eliminates the resistance of the test leads from the measurement since voltage drop is measured directly on the sample, as opposed in the case of two probe measurement. A four-wire connection does not eliminate thermal EMF:



$$R = \frac{(V_1 + V_{th}) - (-V_1 + V_{th} + \delta V_{th})}{|2I|}$$
0 for fast switching, small T change

Current is applied to two adjacent leads using constant current source, and the voltage is measured across the other two leads using digital voltmeter.

The current was then reversed and the voltage measured again. The two values are subtracted, and the difference is divided by two, in order to cancel thermal voltages.

AC bridges use four-wire circuit where <u>thermal EMFs are eliminated</u> Using AC excitation current source. Excitation frequency is fixed or variable. Resistance is measured by passing the selected excitation current through an internal, high-stability reference resistor and sample being measured. Excitation with constant current (small sample resistivity) or constant voltage is possible. It allows for use of sensitive lock in amplifiers with <u>high noise rejection</u>. Introduction to thermal and transport techniques

AC resistance Measurement

Wheatstone bridge not suitable for small resistances. Many bridges are used for thermometry $(10 - 10^{3} \Omega)$, however for highly conductive single crystal samples at low T small noise is needed on signal level ~ $10^{-3} \Omega$.

The LR-700 AC Resistance Bridge uses a 4-wire AC measurement technique. A fixed magnitude AC sine wave current is applied to the sensor resistor through the I-Hi, I-Lo leads. The resulting voltage, returned by the V-Hi and V-Lo leads, is balanced against a known R-Set sine wave (and X-Set cosine wave) resulting in an analog difference signal ΔR and ΔX . The ΔR or ΔX analog signal is continuously sent to the analog temperature control circuitry to generate an analog heater output current. These signals are also available as ±10VDC analog output voltage at rear panel BNC connectors. Meanwhile, a microcontroller calculates values every 200 milliseconds and sends R-Set, ΔR , or R etc. to the display and/or as an output to the IEEE-488 or RS-232 digital interface. When controlling temperature, the temperature setpoint is controlled by R-Set (X-Set). The sensor temperature is indicated by R and the ΔR indicates the difference between the setpoint and sensor temperature. The R-Set sine waves originate in a programmable precision AC attenuator with digital management. This method delivers good linearity, resolution, and near zero AC offset.

The current generator uses reference resistors much greater than the sensor resistance. Johnson noise of the room temperature reference resistor is therefore greatly attenuated when delivered to the sensor by the LR-700 current generator. This technique is superior to wheatstone bridge type circuitry where the sensor is driven by a resistor whose value is comparable to the sensor resistance value.

More on bridges for low noise sample measurement: Physica Scripta 14, 257 (1986)



SENSOR EXCITATION CURRENT TABLE

		100% EXCITATION VOLTAGE, RMS									
		2µV †	6µV †	20µV	60µV	200µV	600µV	2mV	6mV	20mV	
	2mΩ	1ma	3ma	10ma	30ma	LR-710	LR-710	х	х	Х	
	20mΩ	100µa	300µa	1ma	3ma	10ma	30ma	LR-710	LR-710	Х	
R A N G	200mΩ	10µa	30µa	100µa	300µa	1ma	3ma	10ma	30ma	LR-710	
	2Ω	1µa	3µа	10µa	30µa	100µa	300µa	1ma	3ma	10ma	
	20Ω	100na	300na	1µa	Зµа	10µa	30µa	100µa	300µa	1ma	
	200Ω	10na	30na	100na	300na	1µa	Зµа	10µa	30µa	100µa	
	2ΚΩ	1na	3na	10na	30na	100na	300na	1µa	Зµа	10µa	
E	20ΚΩ	0.1na	0.3na	1na	3na	10na	30na	100na	300na	1µa	
	200ΚΩ	LR-740	LR-740	0.1na †	0.3na †	1na	3na	10na	30na	100na	
	2MΩ	LR-740	LR-740	LR-740	LR-740	0.1na †	0.3na †	1na	3na	10na	

Notes: For any particular range and excitation voltage selected, the actual value of excitation voltage appearing across the sensor resistor is directly proportional to the magnitude in ohms of the sensor resistor. † indicates use of variable excitation.

Van der Pauw Method for Resistivity

Sample with thickness d << width and length (plate – like), with no holes. Four ohmic contacts on the sample << dimensions of the sample.



In practice rarely used with single crystal samples since in most cases it is easier to fabricate needle – like sample

Philips Res. Repts. 13, 1 (1958) Philips Res. Repts. 20, 220 (1958)

Crystal Sample Preparation for Resistivity 1

As opposed to heat capacity or to some extent magnetization measurement, electrical transport is easily influenced by extrinsic effects (flux from crystal growth, oxide layer,....)

Sample geometry is obtained using high resolution optical microscope

$$\rho = \frac{V \bullet A}{I \bullet l} = R \frac{A}{l}$$

In polycrystalline samples measuring sample geometry DOES NOT allow determination of ρ due to grain boundaries



Note that in exp. we measure $R= \rho I/A$, Therefore to increase signal/noise ratio, it is good to have long thin samples, especially for highly conductive metals at low T.



CeCoIn₅ four wire resistivity sample for dillution refrigerator measurement. A typical prepared sample size of >>1 mm, 0.2x0.1 mm results in typical sample resistances of < 10 m Ω at low temperatures. Soldering the Ag wires to samples using pure Indium results in contact resistance of ~ 5 m Ω at low T (courtesy of Johnpierre Paglione, UMD) Introduction to thermal and

Four wire resistivity samples - Petrovic lab BNL

Crystal Sample Preparation for Resistivity 2

Flux grown crystals are often easy to prepare, particularly if they have thin needle or plate morphology.

Sometimes, particularly if crystal is grown using other methods, one has to mechanically separate, cut and polish crystals into appropriate geometry since long separation between voltage leads is desirable when measuring low resistances (low T, good conductivity).

Flux grown crystals have often to be etched in acid to remove excess flux that can cause extrinsic effects.

A highly conductive contact between sample and probe wire is required to minimize self-heating effects in low-T resistivity measurements (P=I²R in DC measurement, but also frequency dependence for AC measurement),

Most often employed contact methods are:

- 1. Epo-Tek H20E Ag epoxy cured at T= 100-150 °C
- 2. Silver or carbon paint (dries at room T)
- 3. Gold evaporation of contacts (requires evaporator)
- 4. Soldering pure metals (In, Ga..)
- 5. Soldering metal eutectics:

There is no "best" contact method, some are better at He3 and He4 temperatures, some are stronger, some react with sample..



R (Ω)



Residual Resistivity Ratio

For a simple metal the resistivity can often be modeled by:

 $\rho = f(T) + \rho_0 f(T) \sim AT^2$ at lowest T

and $f(T) \sim BT$ at intermediate T

 ρ_0 reflects finite nature mean free path. In materials synthesis $\rho 0$ can be a measure of crystalline order, however other effects may increase $\rho 0$ (hybridization in heavy fermion systems, various magnetic scattering, etc...

The residual resistivity ratio: RRR, is a measure of how defect free (or defected) a metallic sample is.

RRR = $\rho(\text{high } T) / \rho(\text{low } T)$

Often RRR = $\rho(300 \text{ K})/\rho(2 \text{ K or } 4.2\text{K})$

NOTE: RRR is also R(HT)/R(LT) since factors of l/A cancel out.

High RRR means low ρ_0 .



Introduction to thermal and transport techniques

Phys. Rev. B 66, 054534

Control and Influence of Synthesis Parameters

Influence of synthesis parameters on materials quality Can be estimated using measurement of R and ρ .

In this case influence of Mg stoichiometry, B purity were investigated.

In addition, it was shown that large RRR is direct consequence of sample quality and not extrinsic effect.



Control of Defect Scattering

If the defect density of a compound can be controlled then ρ_0 (and therefore mean free path) can be systematically changed, leading to a constant off-set in resistivity curves. This can be done by annealing of the sample (in some cases).

Superconductivity is easy to spot in r(T) data since it manifests as a sudden drop in r to zero.

(Courtesy of Paul Canfield, (Ameslab Iowa State))



X.Y. Miao et al. / Journal of Alloys and Compounds 338 (2002) 13-19

Hall Effect Measurement

Charge carriers with charge e experience a Lorentz force that produces accumulation:

$$F_B = qvB$$

This produces a Hall field E_H that creates a force opposite to the Lorentz force:

$$F_E = qE_H$$

When two forces are equal, equilibrium is reached so we can determine the drift speed:

$$qvB = qE_H \Longrightarrow v = \frac{E_H}{B}$$

Therefore current density is: $J = nqv = \frac{nqE_H}{B}$ We define Hall coefficient as:

 $R_H = \frac{E_H}{JB} = \frac{1}{nq}$

And Hall mobility of charge carriers: $\mu_H = \sigma |R_H|$:

Introduction to thermal and transport techniques



Experiment measures the Hall voltageV_H = E_H w and the current I = JA = Jwt, so we can write:

$$R_{H} = \frac{E_{H}}{JB} = \frac{V_{H}/w}{(I/wt)B} = \frac{V_{H}t}{IB} = \frac{1}{nq}$$

Since $q = \pm e$ one can estimate carrier concentration n. Sign of $V_H(R_H)$ can tell electron vs. holes

Hall angle:
$$\tan \theta_H = \frac{E_y}{E_x} = \frac{J_x B_z}{E_x nq} = \frac{\sigma_x B_z}{nq} = \frac{q \mu_x B_z}{|q|}$$

Hall Effect in Two Carrier Systems

Carrier system: set of carriers with same mobility, associated with one energy or degenerate levels Hall coefficient can be expressed as:

$$R_{H} = \frac{\rho_{xy}}{H} = \rho_{0} \frac{\alpha_{2} + \beta_{2} H^{2}}{1 + \beta_{3} H^{2}}$$

Where α and β are functions of individual band mobility μ_1, μ_2 and carrier concentrations n_1, n_2 (J. Appl. Phys. 3187 (1999)):





Anomalous Hall Effect in Magnetic Materials



AC Transport option PPMS

Current source and a voltmeter in tandem, with AC bias current 1Hz to 1kHz. Pickup voltage is filtered and only same frequency and form signals are picked up, therefore reducing sources of noise. Four types of measurement: resistivity, Hall effect, I-V curves and critical current.





Samples mounted on ACT puck (Petrovic lab BNL)

Four-Wire Hall Coefficient Measurement (Channel 2) and Resistivity, I-V Trace, and Critical Current Measurement (Channel 1)

Hall Resistivity Sample Mounting in PPMS

Hall voltage is superimposed on bias voltage ($>> V_H$) from two leads which are not perfectly perpendicular, therefore we use 5 wire configuration.

With magnetic field turned off, potentiometer between two leads is used to adjust and null the offset due to sample resistance.

Offset voltage is further reduced by performing negative field sweeps and eliminating parasitic voltage just like thermal emf's.



Introduction to thermal and transport techniques

Effects of Local Moments

Residual resistivity ρ_0 decreases with increased order. Structurally this means less defects. The conduction electrons can also couple to localized magnetic moments, such as on rare earths.

When the moments change from change from gisorder (paramag.) to order (ferromag. antiferromag., or more complex order) there is a scattering decrease a loss of spindisorder scattering. (E Size of ordered ())(Interpreted (Construction))(Interpreted (C degree of hybridization, other scattering sources, etc..!



150

Introduction to thermal and transport techniques

Resistivity of Multiple Transitions

Multiple transitions can be easily detected and identified: In this case we have a rare example of $T_N \sim 10$ K with the loss of spin disorder scattering, followed at lower temperatures by $T_c \sim 6$ K and a total loss of resistivity.



Thermal Conductivity of Electrons

Heat flux due to change of entropy dS: dQ=TdS=dU- μ dN \rightarrow total flux: $J_O = J_U - \mu J_N$ Note single band is involved, sumation is necessary if more $J_U = \frac{1}{4\pi^3} \int \varepsilon(\vec{k}) v(\vec{k}) f(\vec{k}) d\vec{k}; J_N = \frac{1}{4\pi^3} \int (1 \cdot v(\vec{k}) d\vec{k} \Rightarrow J_Q = \frac{1}{4\pi^3} \int (\varepsilon(\vec{k}) - \mu) v(\vec{k}) f(\vec{k}) d\vec{k}$ We again apply relaxation time approximation, just like for σ , but here E=0 and spatial variation of f is due to T gradient $\nabla_r f_0 = \nabla_r T \bullet \partial f_0 / \partial T$ so we have : $f(\vec{k}) \approx f_0(\vec{k}) - \tau(\vec{k})\vec{v}\nabla_r f_0(\vec{k})$ $J_{Q,x} \approx \frac{1}{4\pi^3} \int \left(\varepsilon(\vec{k}) - \mu \right) v_x^2(\vec{k}) \tau(\vec{k}) \frac{\partial f_0}{\partial T} \left(-\frac{\partial T}{\partial x} \right) d\vec{k}$ $\rho(\varepsilon) = \frac{1}{4\pi^3} \int_{\varepsilon(\vec{k})=const} \frac{dS_{\varepsilon}}{\nabla_k \varepsilon}; \langle v_x^2 \rangle = v^2 / 3;$ Let's evaluate this part: states per volume Now $\tau(\varepsilon) = \tau(\varepsilon_F)$, $v = (\varepsilon) = v(\varepsilon_F)$, const. and $\int \varepsilon \rho(\varepsilon) \frac{df_0}{dT} d\varepsilon = \frac{\partial}{\partial T} \int \varepsilon \rho(\varepsilon) f_0(\varepsilon) d\varepsilon = C_v$ We get: $\vec{J}_Q = -\kappa \nabla_r T \Rightarrow \kappa = \frac{1}{2} v_F^2 \tau(\varepsilon_F) C_v$ T dependence governed by C_V and τ .

Thermal Conductivity in Metals



Thermal Conductivity of Phonons



On the "Hot" side phonons with higher energy are excited while on the other side only low energy phonons are excited by establishment of thermal gradient.

The high-energy phonons propagate from left to right at the phonon velocity. Phonon thermal conductivity can be decreased by phonon scattering – dislocations in the crystal lattice, grain boundaries, impurities, etc..

Thermal conductivity in metals has electron and phonon part, in insulators there are no electrons thus thermal conductivity comes from the phonons.

T variation produces non-equilibrium distribution of phonons that propagate with group velocity v_g in a wave packet with $|\Delta k| \sim 1/\Delta x_{Introduction to thermal and transport techniques}$

Thermal Conductivity of Phonons

Harmonic approximation: wave packet will travel without scattering since phonons are independent normal modes $\rightarrow \kappa$ is infinite. However, structural defects, surfaces, grain boundaries..are causes of scattering that changes phonon momentum. We look for heat flux $J_Q - \underline{sum of energies carried by all phonon modes \cdot v_g}$:

$$J_{Q} = \frac{1}{V} \sum_{k,p} n(\vec{k}, p) \hbar \varpi_{k} \vec{v}_{g}(\vec{k}, p) = \frac{1}{V} \sum_{k,p} \hbar \varpi_{k}(p) (n - n^{0})_{k,p} \vec{v}_{g}(\vec{k}, p)$$
In equilibrium no T gradient,

$$J_{Q} = \frac{1}{V} \sum_{k,p} n(\vec{k}, p) \hbar \varpi_{k} \vec{v}_{g}(\vec{k}, p) = \frac{1}{V} \sum_{k,p} \hbar \varpi_{k}(p) (n - n^{0})_{k,p} \vec{v}_{g}(\vec{k}, p)$$
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Thermal Conductivity of Phonons



Thermal Conductivity Example



Mean Free Path and Electron Scattering

Wiedemann Franz law:



Valid in relaxation-time approximation, if some scattering mechanism reduces κ and σ with the same rate, for example Elastic scattering or inelastic where $\Delta \varepsilon << k_B T$. For example at T=0 (where electronic + defect mechanism dominates Or at T>> θ_D where phonons dominate. In intermedate range inelastic e⁻ - ph scattering Causes greater degradation of thermal currents since both electron ε and momentum are changed. Note that at T=0 it does not depend on electronic interactions



Thermal Transport Measurements: Steady vs. Pulse



Figure 2 Schematic representation of experimental set-up: K - thermal conductance, C = thermal capacity, I = current source, T_0 = bath temperature



Figure 3 Time dependence of temeprature difference across the sample: (-----) simulation; (O) experiment

O. Maldonado, Cryogenics 32, 908 (1992)

<u>Steady method</u>: applying heat and measuring temperature (ΔT) and voltage gradients in equillibrium \rightarrow wait times can be big due to <u>T stabilization</u> (that includes offset registration) and <u>establishment of gradient</u>

<u>Pulse method</u>: slow drift in bath T, heat current that generates temperature gradient is pulsed. Resulting signal is exponential for both ΔT and ΔV . We assume $\Delta T = (T_1 - T_0) << (T_1 + T_0)/2 = \overline{T}$

$$\frac{dQ}{dt} = C(T_1) \frac{dT_1}{dt} = I^2(t)R(T_1) - K(T_1 - T_0)$$

If one runs step – function I through heater, after some t equilibrium is reached where $dT_1/dt = 0$ so:

$$K(\overline{T}) = \frac{I^2 R(T_1)}{T_1 - T_0}$$

But if T_0 drifts slowly and if we let periodic square wave Through the heater where I = 0 for $t = [|2\tau < \tau|]$ and $I = I_0$ for $t = [|2\tau \ge \tau|]$, T of the system will be oscillatory with exponential solutions for $\Delta T(t)$. See

Thermal Transport Measurements in PPMS 1

PPMS TTO option enables simultaneous measurement of thermal conductivity κ , Seebeck coefficient S (thermopower) and (AC) resistivity \leftrightarrow enables measurement of $Z = S^2/\kappa\rho$

Continuous measurement: taken continuously and software adjusts parameters (heater power and period) to optimize measurement.

Single measurement: slower, system reaches equilibrium with heater "on" and "off" before taking data point. No curve fitting is required. Less dependence on thermal history

Hot

Measurement of κ : Heater applies heat, and ΔT is measured (W/mK):

$$\frac{\Delta Q}{\Delta t} = \kappa A \frac{\Delta T}{x} \Longrightarrow \kappa = \frac{\Delta Q x}{\Delta t \Delta T A}$$

Q (heat flow)

Measurement of S: Heater applies heat, and ΔV is measured (V/K):

Measurement of ρ : Same as in ACT, same probes (leads) used for thermal and electrical measurements, electrical measurements are taken at the beginning and at the end of the heat pulse



Thermal Transport Measurements in PPMS 2



TTO 3

Continuous measurement: optimization of heater current, heat pulse period and R excitation.



Least square fitting of these variables after ΔT vs time over the <u>duration of the heat pulse</u> data are obtained. Fitting routine searches for T_{∞} , τ_1 and τ_2 :



different from sample S (dominated by τ_1)

<u>Cooling "pulse"</u> is fitted with the profile:

$$\Delta T_{\text{cooling}} = A - \Delta T_{\text{cooling}}$$

Thermal history is accounted in the model by taking into account two previous pulses. Fitting for Seebeck voltage is similar:

Linear (account for varying thermal V) drift and offset voltage

Thermal Transport Measurement in PPMS 4

Errors in the measured quantities are calculated in the standard manner. For ΔT vs time curve it is calculated as: Number of data points in the curve ∇

$$R_{\Delta T} = \sqrt{\frac{\sum_{i} (\Delta T_{i} - \Delta T_{i,\text{model}})^{2}}{N}} + N = 64 \text{ for } \kappa, \text{ S and 128 for } \rho$$

Error in sample radiation, 20% estimated combined error from emissivity and sample radiation
Standard deviation in thermal conductivity is:
$$\sigma(\kappa) = \kappa \sqrt{\left(\frac{R_{\Delta T}}{\Delta T_{\infty}}\right)^{2} + \left(\frac{2IR\delta I}{P}\right)^{2} + \left(\frac{0.2Ploss}{P}\right)^{2} + \left(\frac{0.1T_{\infty}K_{shoes}}{P}\right)^{2}}$$

 $\sqrt{2}$

Voltage error:

1/

 ~ 2

 $\searrow 2$

Resistivity error (average of beginning and end heat pulse measurement:

$$\sigma(\alpha) = \alpha \sqrt{\left(\frac{R_{\Delta V}}{\Delta V_{\infty}}\right)^{2}} + \left(\frac{R_{\Delta T}}{\Delta T_{\infty}}\right)^{2} \qquad \text{Peak to peak V} \qquad \sigma(\rho) = \rho \frac{R_{\rho}}{V_{pp}} = \frac{\rho}{V_{pp}} \sqrt{\frac{\sum_{i} (V_{i} - V_{i}, \text{mod } el)^{2}}{N}}$$

$$ZT \text{ error is simply: } \sigma(ZT) = ZT \sqrt{\left(\frac{2\sigma(S)}{S}\right)^{2} + \left(\frac{\sigma(\kappa)}{\kappa}\right)^{2} + \left(\frac{\sigma(\rho)}{\rho}\right)^{2} + \left(\frac{\sigma(T)}{T}\right)^{2}}$$

 $(I^2 R - (P_{rad}))$ $P_{rad} = \sigma_T (A_{total}/2) c(T_{hot}^4 - T_{cold}^4)$ Heat loss errors: K[W/K] = $= aT + bT^2 + cT^3$) →K_{shoes} $\frac{1}{2}$ of surface area is radiating, the other is cold shoes σT – Stefan – Boltzmann constant 5.67·10-8 (W/K⁴m² Introduction to thermal and transport techniques

 ε – infrared emissivity, important above 300 K since some Radiative heat power is lost To surrounding heat shield

TTO Calibration Files

Relative errors in measurement can be ~ 10 - 20%, above 300K (radiation heating) and when measuring small κ . Note that errors become large near large changes, phase transitions (metal – insulator in this case).



Thermoelectric Materials: FeSb₂

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semiconductor FeSb₂

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2008: 40th Anniversary of the EPS Ethics of nanotechnology or... nanoethics? Forum Physics and Society I & II "On the way we conference" Global mapping of greenhouse gases and air pollutants

Institutional subscri 99 euros per year Colossal Seebeck coefficient in semiconducting FeSb.

Rare-earth, actinide or transition-metal containing correlated semiconductors are characterised by a small hybridisation gap at the Fermi level from mixing of a broad conduction band with a narrow dor f-band. Such materials are expected to have large absolute values of the Seebeck coefficient, S=V/DT, where V is the voltage difference due to charge carrier diffusion from a hot to a cold region of a material with a temperature difference T.

The figure shows the electrical resistivity (ρ) of single crystalline FeSb₂. $\rho(T)$ is semi-conducting and decreases with increasing temperature with a 'shoulder' in the temperature range from 10 K to 30 K. The Seebeck coefficient (S) drops to a sample dependent local minimum which in one sample reaches the record low value of -45000 mVK-1. A plausible interpretation of these properties involves excitations of charge carriers across a very narrow band-gap formed by Fe 3d states weakly hybridised with

Sb 5p states whose coherence starts to deteriorate when the temperature increases above 10 K. The thermoelectric conversion efficiency increases monotonically with $ZT = (S^2 \cdot \mathbf{r} - \rho^{-1}/\kappa) \cdot T$, κ being the thermal conductivity. Materials with ZT > 1 are considered useful for thermoelectric cooling or power applications. The thermoelectric power factor $PF=S^2 \cdot r - \rho^{-1}$, an important part of ZT, is also shown in the figure. At 12 K the power factor reaches a record high value of ~2300 mWK-2 cm-1, however a large κ reduces ZT to 0.005. Nonetheless, if κ is reduced to ~3 Wm⁻¹K⁻¹, as observed in *e.g.* thin films, $ZT \sim 1$ can be obtained thus underlining the potential of FeSb₂ as a future solid-state thermoelectric cooling device.

Anders Bentien, Simon Johnsen, Georg Kent Hellerup Madsen, Bo Brummerstedt Iversen and Frank Steglich,

"Colossal Seebeck coefficient in strongly

correlated semiconductor FeSb₃", Eur. Phys. Lett. 80, 17008 (2007)

Colossal Seebeck coefficient in strongly correlated

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Electrical transport properties as function of temperature along different crystallographic directions on different single crystalline FeSb, samples.

Introduction to thermal and transport techniques

Figure 1 Schematic drawing of a thermoelectric device. The application of a temperature gradient generates a voltage that can be used to power electrical devices.



Introduction to thermal and transport techniques

Thermoelectric Devices

	Wavelenth	Spectrum	%
Photovoltaic	~200–800nm	UV & visible light	58
Thermoelectric	~800–3000nm	IR	42



Figure 2. Sun radiates energy as a 6000 K blackbody radiator with part of the energy in the ultraviolet (UV) spectrum and part in the infrared (IR) spectrum.



Figure 3. After the solar energy is concentrated, it can be converted into electricity through both photovoltaics (UV spectrum) and thermoelectrics (IR spectrum).